

Structured Connectivity for 6G Reflex Arc: Task-Oriented Virtual User and New Uplink-Downlink Tradeoff

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Abstract—To accommodate the evolving demands of unmanned operations, the future sixth-generation (6G) network will support not only communication links but also sensing-communication-computing-control (SC³) loops. In each SC³ cycle, the sensor uploads sensing data to the computing center, and the computing center calculates the control command and sends it to the actuator to take action. To maintain the task-level connections between the sensor-to-computing-center link and the computing-center-to-actuator link, we propose to treat the sensor and actuator as a virtual user. In this way, the two communication links of the SC³ loop become the uplink and downlink (UL&DL) of the virtual user. Based on the virtual user, we propose a task-oriented UL&DL optimization scheme. This scheme jointly optimizes UL&DL transmit power, time, bandwidth, and CPU frequency to minimize the control linear quadratic regulator (LQR) cost. We decouple the complex problem into a convex UL&DL bandwidth allocation problem with the closed-form solution for the optimal time allocation. Simulation results demonstrate that the proposed scheme achieves a task-level balance between the UL&DL, surpassing conventional communication schemes that optimize each link separately.

Index Terms—Joint uplink and downlink (UL&DL) optimization, sensing-communication-computing-control (SC³) loop, task-oriented communication, virtual user.

I. INTRODUCTION

CURRENTLY, wireless communication has expanded from connecting humans to connecting machines. Driven by the increasing demands for unmanned operations, supporting field robots has been identified as an important use case for the sixth-generation (6G) communication [1]. Take disaster rescue as an example. Once the emergency happens, the network swiftly connects sensors, actuators, and the computing center into sensing-communication-computing-control (SC³) loops, which execute various tasks based on the periodical

control. In each SC³ cycle, the sensor uploads the collected data to the computing center, the computing center calculates the control command and sends it to the actuator, and the actuator takes action. Relying on the effective feedback, the SC³ loop dynamically adapts to varying environments and finishes different tasks without human intervention.

As shown in Fig. 1, the SC³ loop bears a striking resemblance to the reflex arc. Dating back to the 16th century, when Descartes conceptualized the reflex arc, biological researchers found that “the presence of reflexes is only dependent on the functional integrity of the components of the reflex arc” [2]. Similarly, the task-execution capability of the SC³ loop is only dependent on its integrity. We need to take it as a basic unit for the task execution. However, the fifth-generation (5G) network takes the communication link as the basic unit [3]. It applies the multiple access technique to separate different users and applies the duplex technique to separate the uplink and downlink (UL&DL). Although this link-level decomposition brings high capacity to the network, it breaks the task-level connections among the SC³ loop components, making it less effective for the task-oriented 6G networks.

To investigate the SC³ loop, the founder of cybernetics, Norbert Wiener, emphasized that “the problems of control engineering and of communication engineering were inseparable, and that they centered not around the technique of electrical engineering but around the much more fundamental notion of the message” [4]. Therefore, when we are concerned about the “fundamental notion of the message”—the information usage behind data transmission, it is easy to find that the two communication links within one SC³ loop are not independent for their own transmission but are interconnected for executing a common task. To maintain the task-level connections between the two links, we propose to regard the sensor and actuator as a virtual user. Consequently, the link between the sensor and the computing center, along with the link between the computing center and the actuator, are considered the UL&DL of the virtual user. This virtual-user-based approach allows us to take the UL&DL as two interconnected links and jointly configure their communication capabilities. Thus, the UL&DL could be aligned to support the functioning of the SC³ loop.

In the literature, few works investigated the joint UL&DL optimization given that frequency division duplexing (FDD) and time division duplexing (TDD) are commonly used to separate UL&DL. In the early stage, El-Hajj et al. addressed the demands for balanced UL&DL and proposed a sum-rate maximization scheme. The UL&DL rate difference was

Received 3 October 2024; accepted 29 October 2024. Date of publication 7 November 2024; date of current version 10 January 2025. This work was supported in part by the National Natural Science Foundation of China under Grant 62425110, Grant 62341110, and Grant U22A2002; in part by the National Key Research and Development Program of China under Grant 2020YFA0711301; in part by the Suzhou Science and Technology Project; and in part by the FAW Jiefang Automotive Company Ltd. The associate editor coordinating the review of this article and approving it for publication was H. Yue. (Corresponding author: Wei Feng.)

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Digital Object Identifier 10.1109/LWC.2024.3493167

considered as a constraint in the optimization [5]. Then, the emerging time-sensitive applications put high requirements on the round-trip time, necessitating the joint UL&DL design [6], [7], [8], [9]. The work in [6], [7], [8] investigated the multi-access edge computing system, and proposed the energy/power minimization schemes under the round-trip time constraints. The work in [9] proposed a bandwidth minimization scheme that optimizes the UL&DL bandwidth and delay components. However, these studies still focused on communication metrics such as spectrum efficiency and transmission latency, while ignored the task performance behind the data transmission. When the focus shifts to the SC^3 loop, in the field of wireless control system (WCS), researchers investigated the similar control closed loop, and they paid attention to the control performance behind communication [10]. For example, Gatsis et al. focused on the uplink (UL) transmission and proposed a power adaptation scheme, which adapts the sensor power to both plant states and channel states to minimize the linear quadratic regulator (LQR) cost [11]. Extending to the multi- SC^3 loops, Chang et al. optimized UL bandwidth, power, and control convergence rate to maximize the spectrum efficiency [12]. Focusing on the downlink (DL) transmission, Fang et al. optimized the transmit power and block length [13], and Ana et al. optimized the bandwidth [14]. These studies provided great insights for the SC^3 -loop design. However, they all focused on one link and assumed the other link is ideal. There lacks the work that takes the SC^3 loop as an integrated structure and jointly configures UL&DL from a task-oriented perspective.

In this letter, we investigate the SC^3 loop and regard the sensor and actuator as a virtual user. Based on the virtual user, we propose a task-oriented joint UL&DL optimization scheme. Using the control LQR cost as the objective, we jointly optimize UL&DL transmit power, time, bandwidth, and computing CPU frequency. This complex problem is then simplified into a convex bandwidth allocation problem with the closed-form solution for the time allocation. The optimal LQR cost is also expressed in a closed form. Our simulation results underscore the superiority of the proposed task-oriented UL&DL optimization scheme, showing that it achieves a task-level balance between UL&DL.

II. VIRTUAL-USER-BASED SC^3 LOOP MODEL

As shown in Fig. 1, the SC^3 loop encompasses three nodes: the sensor, the computing center, and the actuator, connected by two communication links: one between the sensor and the computing center, and another between the computing center and the actuator. To maintain the task-level connections among SC^3 loop components, we regard the sensor and actuator as a single virtual user. In this setup, the sensor utilizes the UL to upload sensing data, while the actuator employs the DL to receive commands. The considered SC^3 loop is executing a control-type task. Without loss of generality, we model the controlled system as a linear time-invariant system. The discrete-time evolution equation is given by,

$$\mathbf{x}_{i+1} = \mathbf{A}\mathbf{x}_i + \mathbf{B}\mathbf{u}_i + \mathbf{v}_i, \quad (1)$$

where i is the time index, $\mathbf{x}_i \in \mathbb{R}^{n \times 1}$ is the system state, $\mathbf{u}_i \in \mathbb{R}^{m \times 1}$ is the control action, $\mathbf{v}_i \in \mathbb{R}^{n \times 1}$ is the process noise, and $\mathbf{A} \in \mathbb{R}^{n \times n}$ and $\mathbf{B} \in \mathbb{R}^{n \times m}$ are determined by the system dynamics. To measure the control performance, the LQR cost is used, which is a weighted summation of the state derivation and control input,

$$l = \limsup_{N \rightarrow \infty} \mathbb{E} \left[\sum_{i=1}^N \left(\mathbf{x}_i^T \mathbf{Q} \mathbf{x}_i + \mathbf{u}_i^T \mathbf{R} \mathbf{u}_i \right) \right], \quad (2)$$

where $\mathbf{Q} \in \mathbb{R}^{n \times n}$ and $\mathbf{R} \in \mathbb{R}^{m \times m}$ are two weight matrices.

Denote UL&DL transmit power, time, and bandwidth as $p_{u/d}$, $t_{u/d}$, and $B_{u/d}$, respectively. For simplicity, we use the subscript u/d to represent UL/DL. Within an SC^3 cycle, the UL&DL cycle rate, denoted as $D_{u/d}$, is given by:

$$D_{u/d} \leq t_{u/d} R(p_{u/d}, B_{u/d}) \quad (\text{bits}/\text{SC}^3 \text{ cycle}), \quad (3)$$

where $R(p_{u/d}, B_{u/d})$ denotes the data rate,

$$R(p_{u/d}, B_{u/d}) \triangleq B_{u/d} \log_2 \left(1 + \frac{p_{u/d} h_{u/d}}{B_{u/d} N_0} \right) \quad (\text{bits/s}), \quad (4)$$

where $h_{u/d}$ denotes the channel gain, which is assumed to be constant along the control process, and N_0 denotes the noise power spectral density. For computing, it is modeled as an information-extraction process, which can be described as,

$$D_u \rightarrow \rho D_u, \quad (5)$$

where ρ ($0 < \rho < 1$) denotes the information extraction ratio, and ρD_u quantifies the task-related information extracted from the sensing data, which is also considered as the volume of information contained in the command. Denote f as the CPU frequency. The computing time is calculated by,

$$t_c = \frac{\alpha D_u}{f}, \quad (6)$$

where α (cycles/bit) denotes the required CPU cycles to process one-bit raw data, which quantifies the computation complexity. Then, the command is sent to the actuator. Considering the constraint of the DL capacity, the task-related information received by the actuator is jointly determined by the UL&DL cycle rate:

$$D_{\text{SC}^3} = \min(\rho D_u, D_d) \quad (\text{bits}/\text{SC}^3 \text{ cycle}). \quad (7)$$

D_{SC^3} quantifies the information that finally takes effect for the control task within an SC^3 cycle, which we refer to as the closed-loop negentropy rate (CNER). Denoting the SC^3 -cycle time as T , we have the following cycle-time constraint:

$$t_u + t_c + t_d \leq T. \quad (8)$$

Moving forward, the lower bound of the LQR cost has the following relationship with the CNER [15]:

$$l \geq \frac{nN(\mathbf{v}) |\det \mathbf{M}|^{\frac{1}{n}}}{2^{\frac{2}{n}} (D_{\text{SC}^3} - \log_2 |\det \mathbf{A}|) - 1} + \text{tr}(\Sigma_{\mathbf{v}} \mathbf{S}), \quad (9)$$

where $N(\mathbf{x}) \triangleq \frac{1}{2\pi e} e^{\frac{2}{n} h(\mathbf{x})}$, $h(\mathbf{x}) \triangleq -\int_{\mathbb{R}^n} f_{\mathbf{x}}(x) \log f_{\mathbf{x}}(x) dx$, $\Sigma_{\mathbf{v}}$ is the covariance matrix of the process noise, $\log_2 |\det \mathbf{A}|$

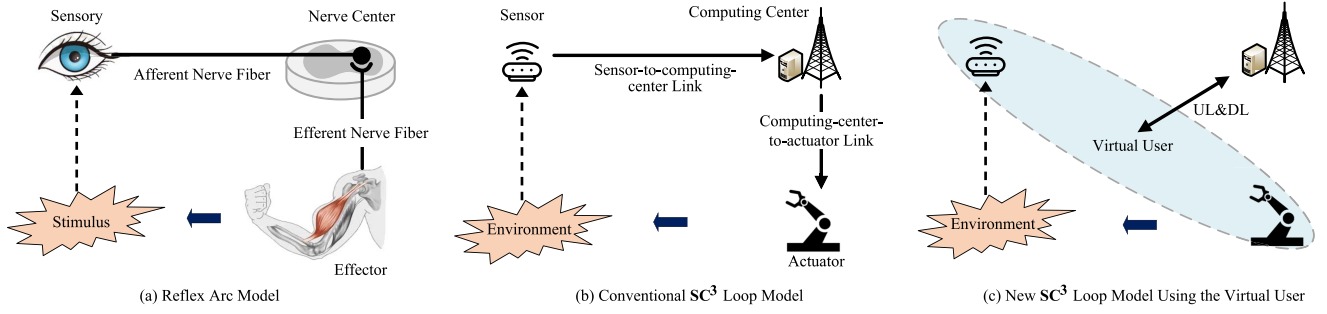


Fig. 1. Illustration of the reflex arc model, conventional SC³ loop model, and new SC³ loop model using the virtual user.

is the intrinsic entropy rate, and $\mathbf{M}$ and $\mathbf{S}$ are solved by the following Riccati equations:

$$\mathbf{S} = \mathbf{Q} + \mathbf{A}^T(\mathbf{S} - \mathbf{M})\mathbf{A}, \quad \mathbf{M} = \mathbf{S}^T \mathbf{B}(\mathbf{R} + \mathbf{B}^T \mathbf{S} \mathbf{B})^{-1} \mathbf{B}^T \mathbf{S}. \quad (10)$$

To ensure the system can be stabilized, the CNER needs to exceed the intrinsic entropy rate,

$$D_{\text{SC}^3} > \log_2 |\det \mathbf{A}|. \quad (11)$$

The above stability condition is both necessary and sufficient. Provided (11) is satisfied, there must exist a code-controller that ensures the asymptotic stability, i.e., $\lim_{i \rightarrow +\infty} \sup \|\mathbf{x}_i\| < \infty$ [16]. As a result, (9) makes sense when the stability condition (11) is satisfied, otherwise, the LQR cost is infinite.

III. TASK-ORIENTED UL&DL OPTIMIZATION SCHEME

In the considered SC³ loop model, the LQR cost is bounded by the CNER, which, in turn, is determined by the UL&DL cycle rate. Additionally, the computing speed influences the UL&DL time and indirectly influences the control performance. Therefore, we jointly optimize the UL&DL transmit power, time, bandwidth, and CPU frequency to minimize the LQR cost,

$$(P1) \quad \min_{p_u, t_u, B_u, f, p_d, t_d, B_d \geq 0} l \quad (12a)$$

$$\text{s.t.} \quad l \geq \frac{nN(\mathbf{v})|\det \mathbf{M}|^{\frac{1}{n}}}{2^{\frac{2}{n}(D_{\text{SC}^3} - \log_2 |\det \mathbf{A}|)} - 1} + \text{tr}(\mathbf{\Sigma}_v \mathbf{S}) \quad (12b)$$

$$D_{\text{SC}^3} > \log_2 |\det \mathbf{A}| \quad (12c)$$

$$D_{\text{SC}^3} \leq \min(\rho D_u, D_d) \quad (12d)$$

$$D_{u/d} \leq t_{u/d} R(p_{u/d}, B_{u/d}) \quad (12e)$$

$$t_u + \frac{\alpha D_u}{f} + t_d \leq T \quad (12f)$$

$$B_u + B_d \leq B_{\max}, p_{u/d} \leq P_{\text{umax}/\text{dmax}}, f \leq f_{\max}, \quad (12g)$$

where (12g) represents the resource constraints on the bandwidth, transmit power, and CPU frequency, with the maximal value denoted by $B_{\max}$, $P_{\text{umax}/\text{dmax}}$, and $f_{\max}$. Given that the right-hand side of (12b) is a decreasing function of D_{SC^3} , minimizing the LQR cost is equivalent to maximizing the CNER. In addition, it is easy to find that the optimal solution necessitates the full utilization of the transmit power and CPU

frequency, i.e., $p_{u/d}^* = P_{\text{umax}/\text{dmax}}$ and $f^* = f_{\max}$. On this basis, we denote $R(B_{u/d}) \triangleq R(p_{u/d}^*, B_{u/d})$. Then, (P1) is simplified into an UL&DL time and bandwidth allocation problem as follows:

$$(P2) \quad \max_{t_u, B_u, t_d, B_d \geq 0} D_{\text{SC}^3} \quad (13a)$$

$$\text{s.t.} \quad D_{\text{SC}^3} \leq \min(\rho D_u, D_d) \quad (13b)$$

$$D_{u/d} \leq t_{u/d} R(B_{u/d}) \quad (13c)$$

$$t_u + \frac{\alpha D_u}{f_{\max}} + t_d \leq T \quad (13d)$$

$$B_u + B_d \leq B, \quad (13e)$$

where we omit the stability condition (12c) and test it after solving (P2). If the optimal CNER, $D_{\text{SC}^3}^*$, satisfies the stability condition (11), the LQR cost is calculated by (9). Otherwise, the system cannot be stabilized and the LQR cost is infinite.

Lemma 1: The optimal solution to (P2) is to achieve a task-level balance between UL&DL:

$$\rho D_u^* = D_d^*. \quad (14)$$

Proof: To maximize the CNER, (13c) must be satisfied as an equality at the optimal solution: $D_{u/d}^* = t_{u/d} R(B_{u/d})$. If (14) does not hold at the optimal solution, e.g., $\rho D_u^* > D_d^*$, the optimal CNER is determined by $D_{\text{SC}^3}^* = D_d^*$. However, $D_{\text{SC}^3}^*$ can be further increased by reallocating the UL time to the DL: $t_d R(B_d) \uparrow \rightarrow D_d^* \uparrow \rightarrow D_{\text{SC}^3}^* \uparrow$, as long as $\rho D_u^* > D_d^*$ holds. This contradicts to the assumption that $D_{\text{SC}^3}^*$ is optimal. We can use the similar reasoning to falsify $\rho D_u^* < D_d^*$. Therefore, $\rho D_u^* = D_d^*$ must hold at the optimal solution. ■

By leveraging the task-level balance between UL&DL, we show that (P2) can be converted into a convex bandwidth allocation problem with the closed-form solution for the time allocation.

Theorem 1: The optimal UL&DL bandwidth allocation is achieved by solving the following problem:

$$(P3) \quad \min_{B_u, B_d \geq 0} \frac{1}{\rho R(B_u)} + \frac{1}{R(B_d)} \quad (15a)$$

$$\text{s.t.} \quad B_u + B_d \leq B_{\max}. \quad (15b)$$

The optimal UL&DL time are given by,

$$t_u^* = \frac{\frac{1}{\rho R_u^*} T}{\frac{1}{\rho R_u^*} + \frac{1}{R_d^*} + \frac{\alpha}{\rho f_{\max}}}, \quad t_d^* = \frac{\frac{1}{R_d^*} T}{\frac{1}{\rho R_u^*} + \frac{1}{R_d^*} + \frac{\alpha}{\rho f_{\max}}}. \quad (16)$$

Algorithm 1 The Proposed Optimization Algorithm

Input: Control-related parameters: $n, m, \log_2 |\det \mathbf{A}|, \mathbf{B}, \mathbf{Q}, \mathbf{R}, T$, and Σ_v ; Communication-related parameters: $P_{\text{umax}}, P_{\text{dmax}}, B_{\text{max}}, h_u, h_d$, and N_0 ; Computing-related parameters: α, ρ , and f_{max} ;

- 1: Calculate $\mathbf{S}$ and $\mathbf{M}$ according to (10);
- 2: Calculate the optimal UL&DL transmit power and CPU frequency: $p_{u/d}^* = P_{\text{umax}}/d_{\text{max}}$ and $f^* = f_{\text{max}}$;
- 3: Solve (P3) to obtain the optimal bandwidth allocation, $B_{u/d}^*$;
- 4: Calculate $R_{u/d}^*$ according to (4), and calculate the optimal time allocation according to (16);
- 5: Calculate the optimal CNER, $D_{\text{SC}^3}^*$, according to (17);
- 6: Judge the stability condition (11) and calculate the limit LQR cost, l^* , according to (18).

where $R_{u/d}^* \triangleq R(p_{u/d}^*, B_{u/d}^*)$ and $B_{u/d}^*$ is the solution to (P3). The optimal CNER and LQR cost are given by,

$$D_{\text{SC}^3}^* = \frac{T}{\frac{1}{\rho R_u^*} + \frac{1}{R_d^*} + \frac{\alpha}{\rho f_{\text{max}}}}, \quad (17)$$

$$l^* = \begin{cases} +\infty, & D_{\text{SC}^3}^* \leq \log_2 |\det \mathbf{A}|, \\ \frac{nN(\mathbf{v})|\det \mathbf{M}|^{\frac{1}{n}}}{2^{\frac{2}{n}}(D_{\text{SC}^3}^* - \log_2 |\det \mathbf{A}|) - 1} + \text{tr}(\Sigma_v \mathbf{S}), & \text{otherwise.} \end{cases} \quad (18)$$

Proof: According to (13c) and (14), the optimal DL time can be expressed as a function of the UL time:

$$\rho t_u^* R(B_u) = t_d^* R(B_d) \Rightarrow t_d^* = \frac{\rho t_u^* R(B_u)}{R(B_d)}. \quad (19)$$

The computing time can also be expressed as a function of the UL time,

$$t_c^* = \frac{\alpha D_u^*}{f_{\text{max}}} = \frac{\alpha t_u^* R(B_u)}{f_{\text{max}}}. \quad (20)$$

By substituting (19) and (20) into (13d), the optimal UL time is expressed as a function of the bandwidth:

$$t_u^* + t_c^* + t_d^* = T \Rightarrow t_u^* = \frac{\frac{1}{\rho R(B_u)} T}{\frac{1}{\rho R(B_u)} + \frac{1}{R(B_d)} + \frac{\alpha}{\rho f_{\text{max}}}}. \quad (21)$$

On this basis, the optimal CNER can be expressed as the function of the bandwidth:

$$D_{\text{SC}^3}^* = \rho D_u^* = \rho t_u^* R(B_u) = \frac{T}{\frac{1}{\rho R(B_u)} + \frac{1}{R(B_d)} + \frac{\alpha}{\rho f_{\text{max}}}}. \quad (22)$$

From (22), we can find that maximizing the CNER is to minimize $[\frac{1}{\rho R(B_u)} + \frac{1}{R(B_d)}]$. Thereby, we get (P3). By solving (P3) and substituting the optimal UL&DL bandwidth, B_u^* and B_d^* , into the (22) and (21), we get t_u^* and $D_{\text{SC}^3}^*$. We further get t_d^* by substituting $B_{u/d}^*$ and t_u^* into (19), and get l^* by testing the stability condition (11) and substituting $D_{\text{SC}^3}^*$ into (9). ■

In fact, (P3) is a time-minimization problem. Its objective, $[\frac{1}{\rho R(B_u)} + \frac{1}{R(B_d)}]$, is the UL&DL time for transmitting one-bit task-related information. Therefore, the equivalence between (P2) and (P3) becomes obvious. We summarize the proposed scheme in **Algorithm 1**.

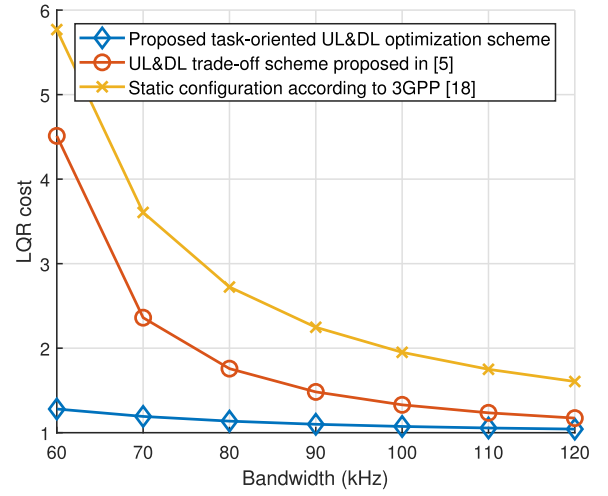


Fig. 2. The LQR cost varies with the available bandwidth resources under three UL&DL configuration schemes.

IV. SIMULATION RESULTS AND DISCUSSION

In this section, we present simulation results. Simulation parameters are set as: $P_{\text{umax}} = 0.1$ W, $P_{\text{dmax}} = 1$ W, $B_{\text{max}} = 1$ MHz, $N_0 = -174$ dBm, and the channel gain are calculated by the path-loss model of $h_{u/d}(\text{dB}) = [32.4 + 20 * \log_{10}(f_c) + 20 * \log_{10}(d_{u/d})]$ [17], where f_c (MHz) represents the carrier frequency and $d_{u/d}$ (km) represents the transmission distance, i.e., $f_c = 2000$ MHz and $d_{u/d} = 1$ km. Control-related parameters are given by $n = m = 100$, $\log_2 |\det \mathbf{A}| = 50$, $\mathbf{R} = \mathbf{0}_{100}$, $\mathbf{Q} = \mathbf{I}_{100}$, $\Sigma_v = 0.01 * \mathbf{I}_{100}$, and $T = 20$ ms [13]. Computing-related parameters are given by $f_{\text{max}} = 1$ GHz, $\alpha = 100$ cycles/bit, and $\rho = 0.01$.

In Fig. 2, we compare the proposed scheme with the UL&DL trade-off scheme proposed in [5] and the static configuration. For fair comparison, we replicate [5] by adopting the sum cycle rate of UL&DL as the objective, $[D_u + D_d]$, and constraining the UL&DL disparities using the constraint $|\rho D_u - D_d| \leq D_0$, where $D_0 = 100$ (bits/SC³ cycle). For the static setup, we apply the equal bandwidth division based on FDD standard and set $t_u = \frac{6T}{7}$, $t_c = \frac{T}{14}$, and $t_d = \frac{T}{14}$, which aligns with the slot format 34 defined in the 5G New Radio (NR) TDD standard [18]. We can see that the proposed scheme consistently exhibits the lowest LQR cost compared with the other two schemes. This outcome highlights the superiority of jointly configuring UL&DL from a task-oriented perspective.

We further present Fig. 3 to reveal the reason behind the superiority of the proposed scheme. In this simulation, the maximum bandwidth is set as $B_{\text{max}} = 1$ MHz. Fig. 3 is a dual-axis chart, with the left axis representing the task-related cycle rate, i.e., $(\rho D_u, D_d)$. From the bars, we can see that under the proposed scheme, the UL&DL are aligned to have the same task-related cycle rate, while the other two schemes fail to achieve this balance. The scheme in [5] allocates more resources to the UL, while the static configuration [18] allocates more resources to the DL. This imbalance results in the DL under the scheme in [5] and the UL under the static configuration [18] becoming the bottleneck, limiting the control performance of the SC³ loop. As a result, the LQR

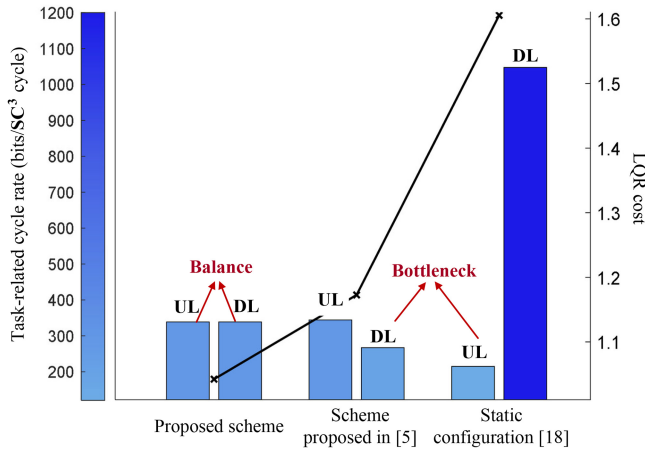


Fig. 3. The UL&DL task-related cycle rate and LQR cost under the proposed scheme, the scheme proposed in [5], and the static configuration [18].

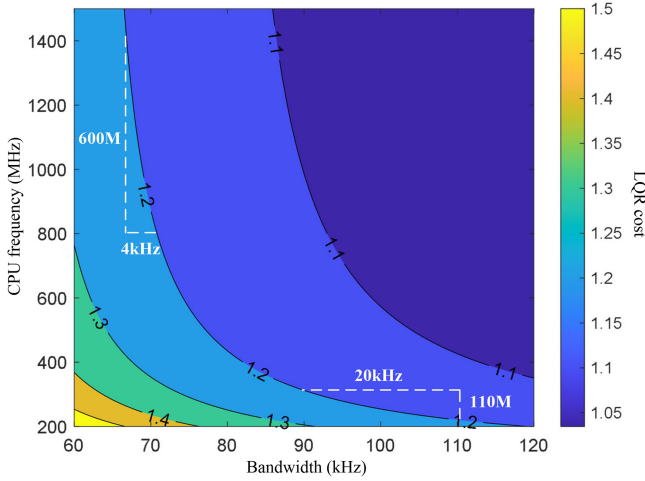


Fig. 4. The LQR cost contour map varying with the bandwidth and computing CPU frequency.

costs (black curve, measured by the right axis) under these two schemes are higher than the proposed scheme. A greater imbalance between UL&DL leads to a higher LQR cost. This verifies the importance of maintaining a task-level UL&DL balance within the SC^3 loop.

Fig. 4 depicts the LQR cost contour map under the optimal UL&DL and computing configuration. We can see that the LQR cost decreases with the enhancement of both computing and communication resources. Furthermore, the map reveals a notable trade-off between communication bandwidth and computing CPU frequency. Take the contour line with the LQR cost of 1.2 as an example. In the bandwidth-constrained region, a 4 kHz increase in bandwidth can offset the need for 600 MHz CPU frequency. Conversely, in the CPU-constrained region, an additional 20 kHz bandwidth only compensates for 110 MHz CPU frequency. It is interesting to find the marginal utility balance between communication and computing such that the operational cost of the SC^3 loop can be minimized.

V. CONCLUSION

In this letter, we have investigated the basic model of the reflex-arc-like SC^3 loop. To maintain the task-level connections between two communication links within the SC^3 loop, we have treated the sensor and actuator as a virtual user and jointly optimized UL&DL transmit power, time, bandwidth, and computing CPU frequency to minimize the LQR cost. We have simplified the complex problem into a convex bandwidth allocation problem, along with the optimal closed-form solution for the time allocation. Our simulation results have confirmed the superiority of the proposed task-oriented UL&DL optimization scheme, highlighting the importance of keeping task-level UL&DL balance within the SC^3 loop.

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